



Fourier Transform of Signals

訊號的傅立葉轉換 Lecture 3-5

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Continuous-Time Non-Periodic Signals The Fourier Transform (FT)

- 連續時間的非週期性訊號表示為複數弦波的疊加
- 複數弦波所包含的頻率是從 $-\infty$ 到 $+\infty$ 連續分佈

$X(j\omega)$ is continuous spectrum and non-periodic.

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{j\omega t} d\omega \quad (7)$$

$x(t)$ is continuous signal and non-periodic.

$$X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt \quad (8)$$

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回憶 FS (if $T \rightarrow \infty$)

$$x(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j\omega_0 k t}$$

$$X[k] = \frac{1}{T} \int_{-T/2}^{T/2} x(t) e^{-jk\omega_0 t} dt$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

→ FT 討論?

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

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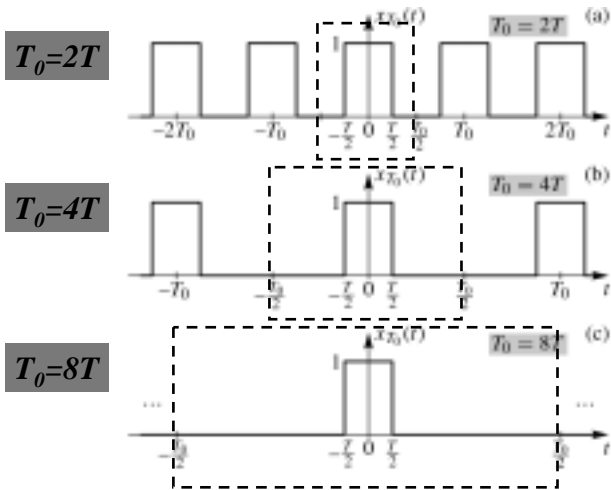


What if Signal $x(t)$ is not periodic?

- Still Sum of Sinusoids?
 - Non-harmonically related sinusoids
 - Would not be periodic, but would probably be non-zero for all t .
- Need “Fourier Transform” to Find Spectrum?
 - gives a “sum” (actually an **integral**) that involves **ALL** frequencies
 - can represent signals that are identically zero for negative t . !!!!!!!!

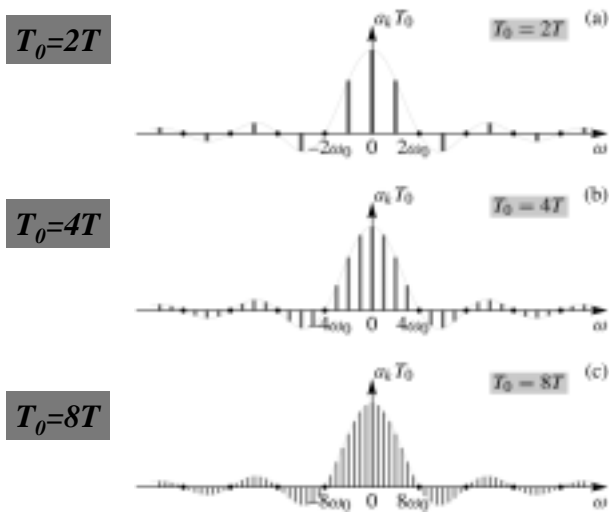
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Limiting Behavior of FS



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Limiting Behavior of Spectrum



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FS in the LIMIT (long period)

Fourier Series Formula:

$$x_{T_0}(t) = \sum_{k=-\infty}^{\infty} X[k] e^{j\omega_0 k t}$$

$k\omega_0 = \omega$ variable

$$2\pi / T_0 = \omega_0$$

$$\lim_{T_0 \rightarrow \infty} 2\pi / T_0 = d\omega$$

$$\lim_{T_0 \rightarrow \infty} x_{T_0}(t) = x(t) \quad \text{for } -\infty < t < \infty$$

$$x_{T_0}(t) = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} [T_0 X[k]] e^{j(\omega_0 k) t} \left(\frac{2\pi}{T_0} \right)$$

$$\Updownarrow$$

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} [X(j\omega)] e^{j(\omega) t} d\omega$$

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Fourier Synthesis

$$x_{T_0}(t) = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} (T_0 X[k]) e^{j\omega_0 k t} \left(\frac{2\pi}{T_0} \right) \mapsto x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

$$\lim_{T_0 \rightarrow \infty} \frac{2\pi}{T_0} = d\omega$$

$$\lim_{T_0 \rightarrow \infty} \frac{2\pi}{T_0} k = \omega$$

$$\lim_{T_0 \rightarrow \infty} T_0 \cdot X[k] = X(j\omega)$$

$$X[k] = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x_{T_0}(t) e^{-j\omega_0 k t} dt \mapsto X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

FS

Fourier Analysis

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Fourier Transform Defined

- For non-periodic signals

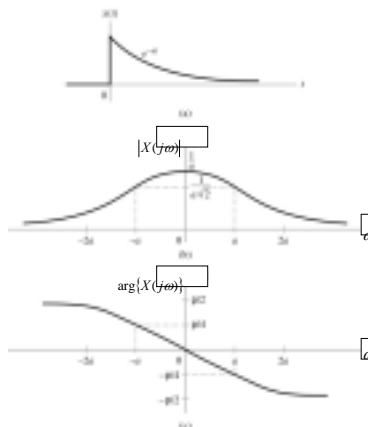
$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} X(j\omega) e^{j\omega t} d\omega$$

Fourier Synthesis

$$X(j\omega) = \int_{-\infty}^{\infty} x(t) e^{-j\omega t} dt$$

Fourier Analysis

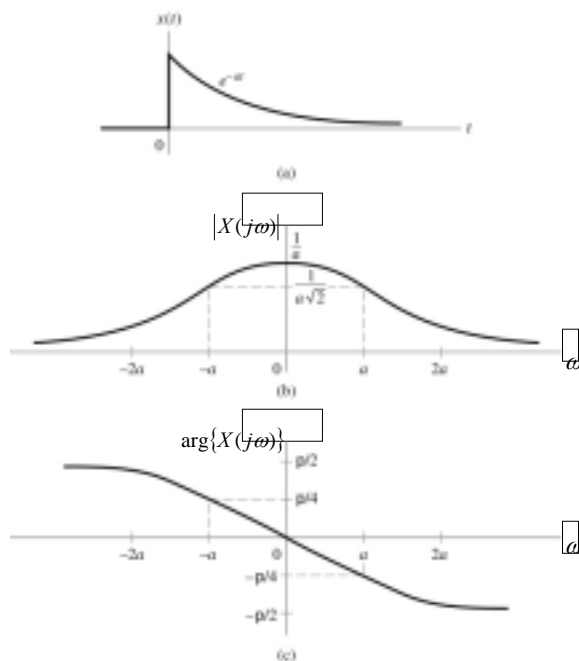
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Ex 3.24 Find the FT of $x(t) = e^{-at}u(t)$:

- Real time-domain exponential signal.
- Magnitude spectrum.
- Phase spectrum.

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Solution of Ex 3.24:

case 1: $a \leq 0$, (Growing Exponential Signal)

$$\int_0^{\infty} e^{-at} dt = \infty. \quad \text{not absolutely summable.}$$

case 2: $a > 0$, (Decaying Exponential Signal)

$$X(j\omega) = \int_{-\infty}^{\infty} e^{-at} u(t) e^{-j\omega t} dt$$

$$= \int_0^{\infty} e^{-(a+j\omega)t} dt = \frac{-1}{a+j\omega} e^{-(a+j\omega)t} \Big|_0^{\infty} = \frac{-1}{a+j\omega} (0-1)$$

$$= \frac{1}{a+j\omega}$$

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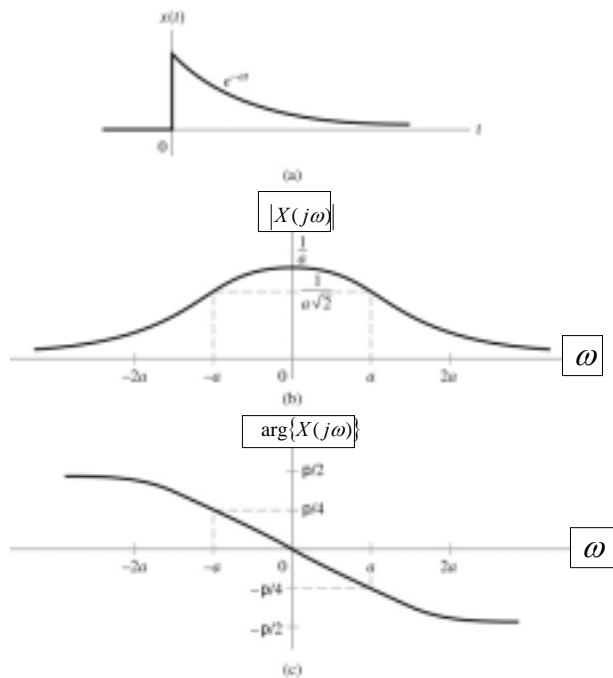


Solution: (cont.)

$$X(j\omega) = \frac{1}{a + j\omega}$$

$$|X(j\omega)| = \frac{1}{\sqrt{a^2 + \omega^2}}; \quad \arg\{X(j\omega)\} = -\tan^{-1}\left(\frac{\omega}{a}\right)$$

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Example 3.25:

Consider the rectangular pulse in the figure (a), and find the FT of $x(t)$, $X(j\omega) = ?$

$$x(t) = \begin{cases} 1, & -T_0 < t < T_0 \\ 0, & |t| > T_0 \end{cases}$$

- (a) Rectangular pulse in the time domain.
 (b) FT in the frequency domain.



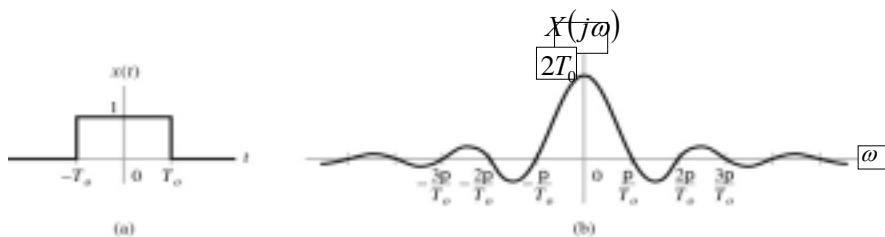
Solution of Ex 3.25:

$$\begin{aligned} X(j\omega) &= \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt \\ &= \int_{-T_0}^{+T_0} 1 \cdot e^{-j\omega t} dt = -\frac{1}{j\omega} e^{-j\omega t} \Big|_{-T_0}^{T_0} \\ &= -\frac{1}{j\omega} (e^{-j\omega T_0} - e^{j\omega T_0}) = \frac{1}{j\omega} (e^{j\omega T_0} - e^{-j\omega T_0}) \\ &= \frac{2}{\omega} \left(\frac{e^{j\omega T_0} - e^{-j\omega T_0}}{j2} \right) = \frac{2}{\omega} \sin(\omega T_0), \quad \forall \omega \neq 0 \end{aligned}$$



$$\forall \omega = 0,$$

$$\lim_{\omega \rightarrow 0} X(j\omega) = \lim_{\omega \rightarrow 0} \frac{2}{\omega} \sin(\omega T_0) = \frac{2}{\omega} (\omega T_0) = 2 T_0$$

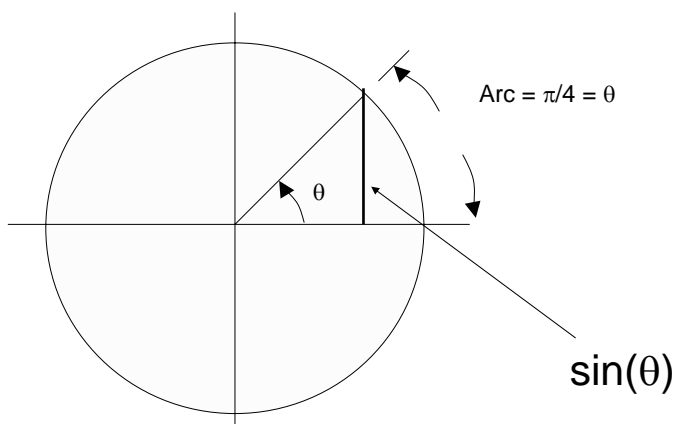


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參考資料

- $\sin(\theta) \approx \theta$ if θ is small.

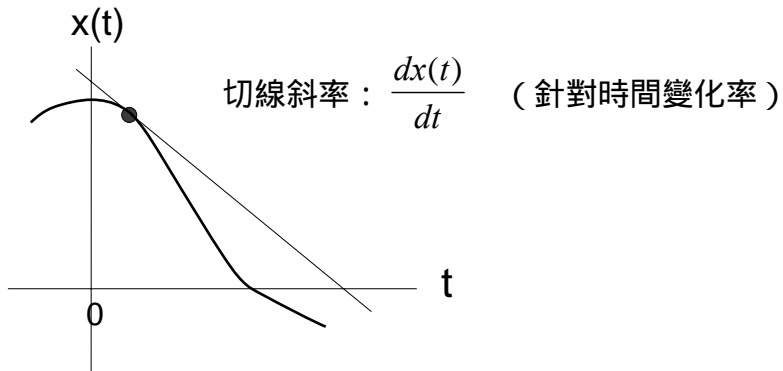


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參考資料

- Changing Rate



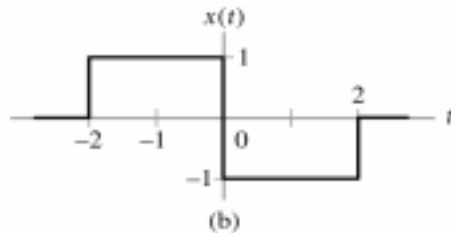
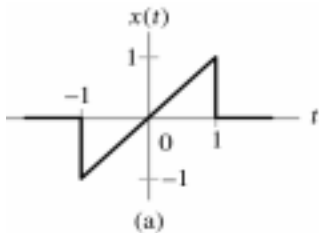
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Time-domain signals for Problem 3.14.

Find the FT of the parts (a) and (b) = ?

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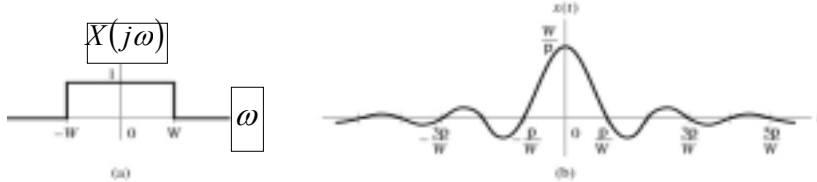
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Example 3.26:

- (a) Rectangular spectrum in the frequency domain.
 (b) Inverse FT in the time domain.

$$X(j\omega) = \begin{cases} 1, & -W < \omega < W \\ 0, & |\omega| > W \end{cases}$$



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Solution of the Ex 3.26:

$$\begin{aligned} x(t) &= \frac{1}{2\pi} \int_{-\pi}^{\pi} X(j\omega) e^{j\omega t} d\omega \\ &= \frac{1}{2\pi} \int_{-W}^W e^{j\omega t} d\omega \\ &= \frac{1}{2\pi jt} \left. e^{j\omega t} \right|_{-W}^W = \frac{1}{2\pi jt} (e^{jWt} - e^{-jWt}) \\ &= \frac{1}{\pi t} \frac{e^{jWt} - e^{-jWt}}{j2} \\ &= \frac{1}{\pi t} \sin(Wt) \end{aligned}$$

$$\text{sinc}(W) = \frac{\sin(\pi W)}{\pi W}$$

$$\frac{W}{\pi} \text{sinc}\left(\frac{Wt}{\pi}\right) = \frac{W}{\pi} \frac{\sin(Wt)}{Wt} = \frac{\sin(Wt)}{\pi t}$$

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Example 3.27: Find the FT of $x(t) = \delta(t)$

Solution:

$$X(j\omega) = \int_{-\infty}^{+\infty} \delta(t) e^{-j\omega t} dt = 1$$

notation :

$$X(j\omega) = FT\{\delta(t)\} = 1$$

or

$$\delta(t) \xleftrightarrow{FT} 1$$

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Example 3.28: Find the inv FT of $X(j\omega) = 2\pi\delta(\omega)$?

Solution:

$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} 2\pi\delta(\omega) e^{j\omega t} d\omega = 1$$

notation :

意義：直流訊號頻率為0

$$x(t) = FT^{-1}\{2\pi\delta(\omega)\} = 1$$

or

$$1 \xleftrightarrow{FT} 2\pi\delta(\omega)$$

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Frequency-domain signals for Problem 3.15.

Find the inverse of FT = ?

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