



Lecture 2-3

Linear Time-Invariant System (LTI System)

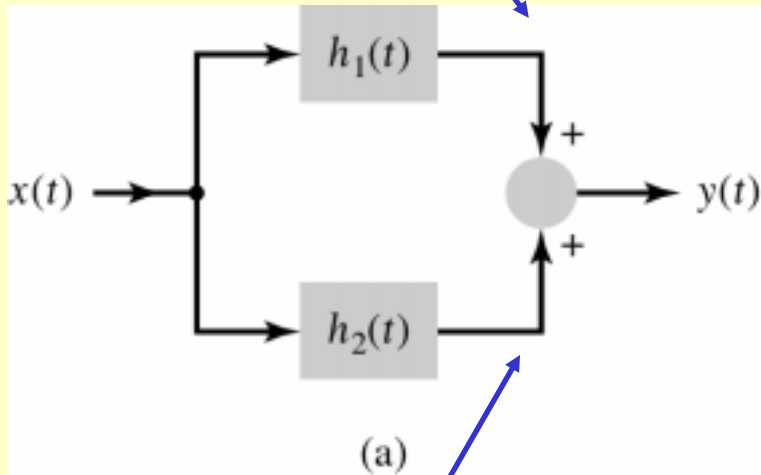
線性非時變系統



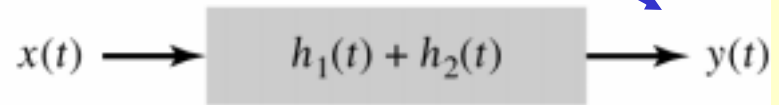
Interconnections of LTI Systems

Interconnection of two LTI systems. (a) Parallel connection of two systems. (b) Equivalent system.

$$x(t) * h_1(t) = \int x(\tau)h_1(t - \tau) d\tau$$



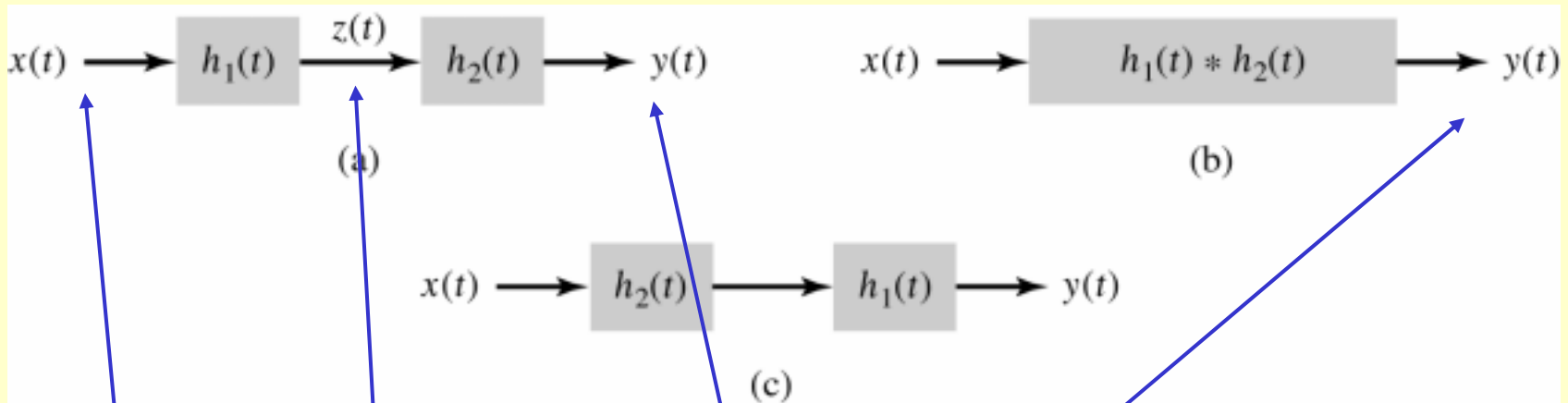
$$\begin{aligned} x(t) * (h_1(t) + h_2(t)) &= \int x(\tau)(h_1(t - \tau) + h_2(t - \tau)) d\tau \\ &= \int (x(\tau)h_1(t - \tau) + x(\tau)h_2(t - \tau)) d\tau \\ &= \int x(\tau)h_1(t - \tau) d\tau + \int x(\tau)h_2(t - \tau) d\tau \end{aligned}$$



$$x(t) * h_2(t) = \int x(\tau)h_2(t - \tau) d\tau$$



Interconnection of two LTI systems. (a) Cascade connection of two systems. (b) Equivalent system. (c) Equivalent system: Interchange system order.

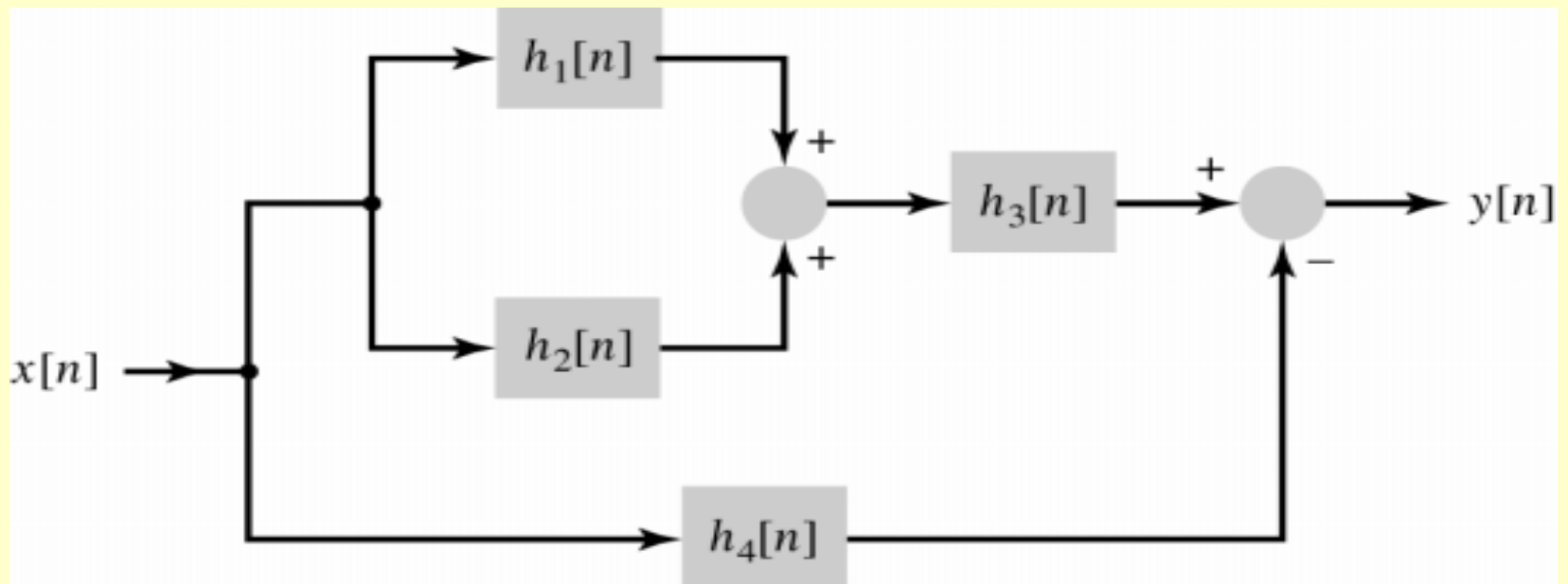


$$\text{if } x(t) = \delta(t), \quad z(t) = h_1(t)$$

$$\therefore y(t) = h_{all}(t) = z(t) * h_2(t) = h_1(t) * h_2(t)$$

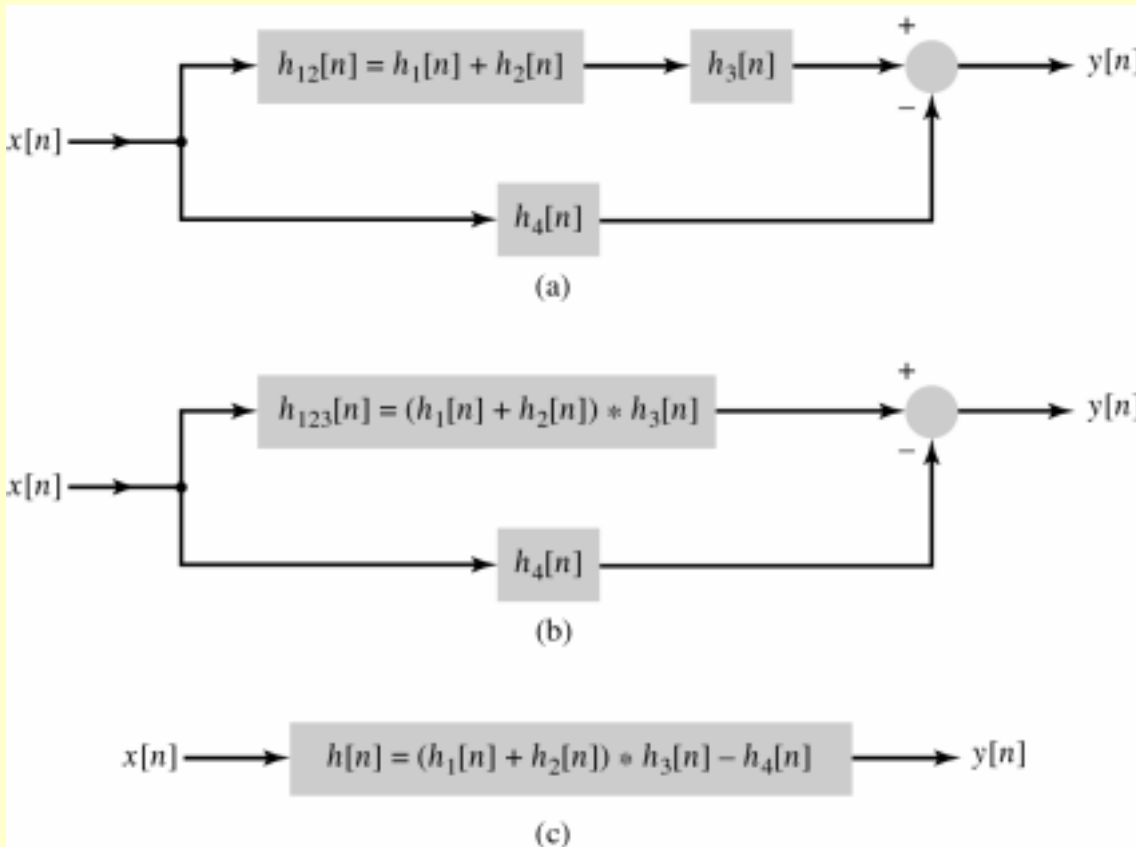


Given impulse responses of each subsystem for the interconnection of systems for Example 2.11, find the total system output =?





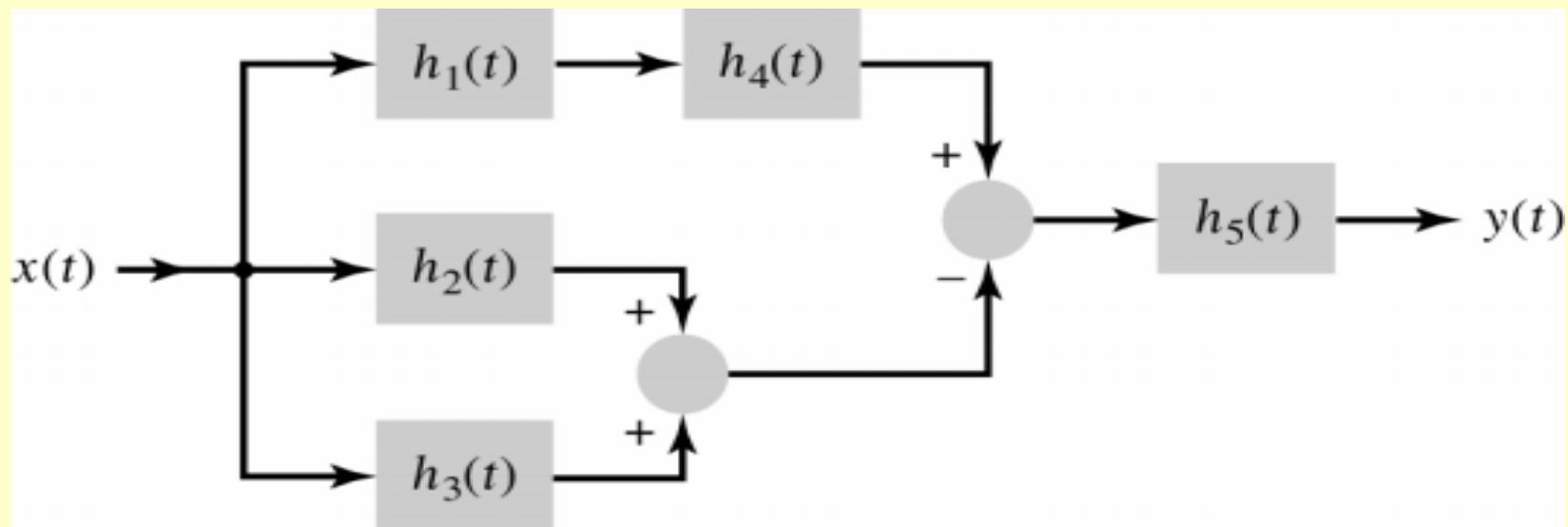
- (a) Reduction of parallel combination of LTI systems in upper branch of Fig. 2.20.
- (b) Reduction of cascade of systems in upper branch of Fig. 2.21(a).
- (c) Reduction of parallel combination of systems in Fig. 2.21(b) to obtain an equivalent system for Fig. 2.20.





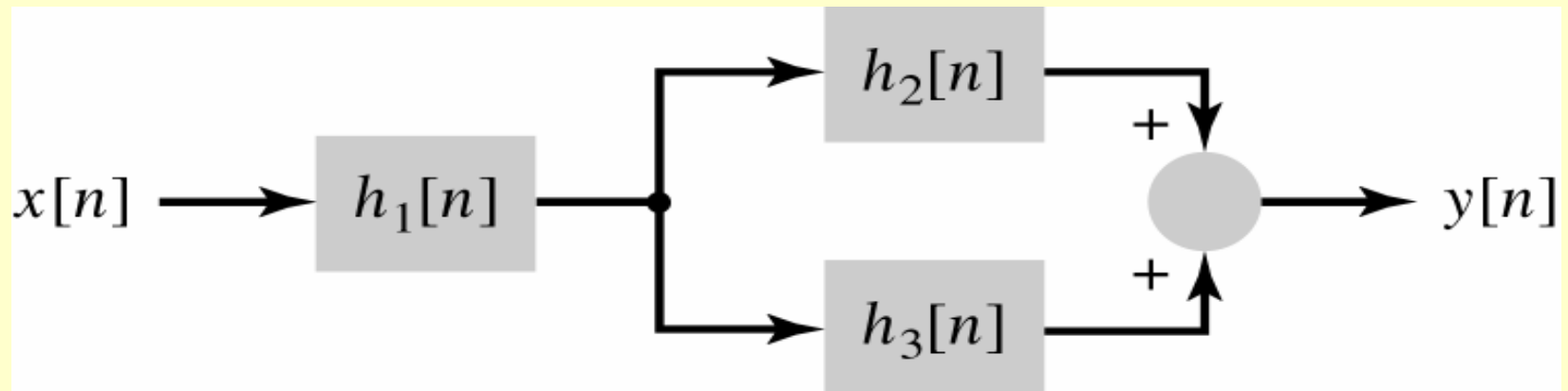
Interconnection of LTI systems for Problem 2.8.

$$h_{all}(t) = h_5(t) * (h_1(t) * h_4(t) - (h_2(t) + h_3(t)))$$





Interconnection of LTI systems for Problem 2.9.





LTI System Properties

- Memory-less LTI System
- Causal LTI System
- Stable LTI System
- Invertible LTI System and De-convolution



Memory-less LTI System: 輸出只與現在輸入有關，輸出為輸入乘以一個純量。

$$\begin{aligned}y[n] &= x[n] * h[n] = h[n] * x[n] \\ &= \sum_{k=-\infty}^{+\infty} h[k]x[n-k] = h[0]x[n],\end{aligned}$$

if and only if $h[k] = c\delta[k]$.

$$y(t) = \int_{-\infty}^{+\infty} h(\tau)x(t-\tau)d\tau = h(0)x(t),$$

if and only if $h(\tau) = c\delta(\tau)$.



Discrete-Time Causal LTI System:

輸出只與現在與過去的輸入有關，系統在脈衝輸入之前不能有輸出。

$$y[n] = h[n] * x[n] = \sum_{k=-\infty}^{+\infty} h[k]x[n-k] \quad \text{or}$$

$$y[n] = \cdots + \underbrace{h[-2]x[n+2] + h[-1]x[n+1] + h[0]x[n]} + h[1]x[n-1] + h[2]x[n-2] + \cdots$$

因為輸出只與現在與過去的輸入有關，令

$$h[k] = 0, \quad \text{for } k < 0$$



Discrete-Time Causal LTI System:

$$y[n] = h[n] * x[n] = \sum_{k=0}^{+\infty} h[k]x[n-k]$$



Continuous-Time Causal LTI System:

$$h(\tau) = 0, \quad \text{for } \tau < 0$$

$$y(t) = h(t) * x(t) = \int_0^{\infty} h(\tau) x(t - \tau) d\tau$$



Discrete-Time Stable LTI System:

$$\therefore |y[n]| = \left| \sum_{k=-\infty}^{+\infty} h[k]x[n-k] \right| \leq \sum_{k=-\infty}^{+\infty} |h[k]| |x[n-k]|$$

If the input is bounded (BI), $|x[n]| \leq M_x < \infty$,

it implies that $|x[n-k]| \leq M_x < \infty$.

Hence,
$$|y[n]| \leq M_x \sum_{k=-\infty}^{+\infty} |h[k]|$$



If the impulse response is absolutely summable,

$$\sum_{k=-\infty}^{+\infty} |h[k]| < \infty$$

Hence, $|y[n]| \leq M_x \sum_{k=-\infty}^{+\infty} |h[k]| < \infty$



Continuous-Time Stable LTI System:

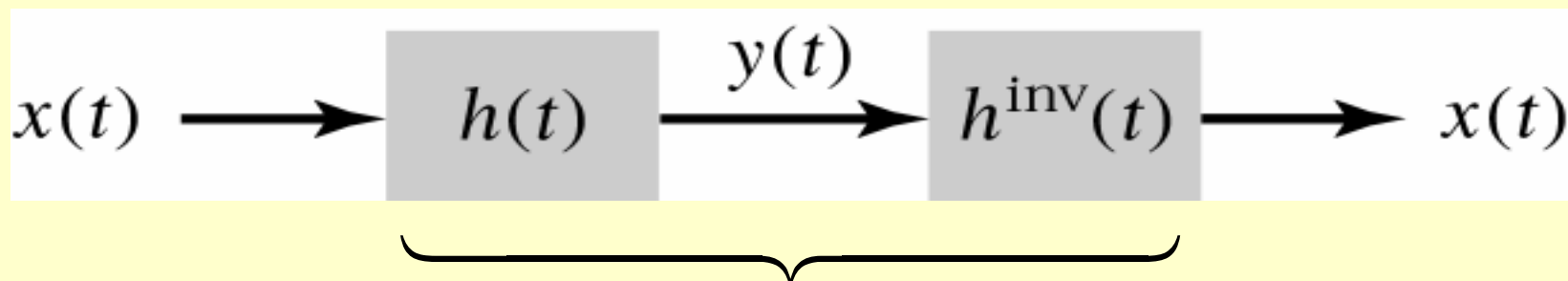
連續時間 LTI 系統是穩定(BIBO) , 若且唯若其脈衝響應是絕對可積分的。 (if and only if)

Iff:

$$\int_{-\infty}^{+\infty} |h(\tau)| d\tau < \infty$$



Invertible System and Deconvolution



$$h(t) * h^{inv}(t) = \delta(t)$$

回復為 $x(t)$ 的過程稱為：Deconvolution



Invertible System and Deconvolution

Continuous-Time:

$$x(t) = x(t) * \{h(t) * h^{inv}(t)\},$$

$$\text{implies, } h(t) * h^{inv}(t) = \delta(t)$$

Discrete-Time:

$$x[n] = x[n] * \{h[n] * h^{inv}[n]\},$$

$$\text{implies, } h[n] * h^{inv}[n] = \delta[n]$$



EX 2.13 $y[n] = x[n] + a x[n - 1]$

Impulse Response $h[n] = ?$

$$h[n] = \delta[n] + a \delta[n - 1]$$

How to find the impulse response from the inverse system, $h^{inv}[n] = ?$

$$h[n] * h^{inv}[n] = \delta[n]$$



$$\because h[n] * h^{inv}[n] = \delta[n]$$

$$\begin{aligned} \therefore h[n] * h^{inv}[n] &= \{\delta[n] + a \delta[n-1]\} * h^{inv}[n] \\ &= h^{inv}[n] + a h^{inv}[n-1] = \delta[n] \end{aligned}$$

when $n = 0$, $h^{inv}[0] + a h^{inv}[-1] = 1$

If the $h^{inv}[n]$ is causal, $h^{inv}[-1] = 0$,

$$\therefore h^{inv}[0] = 1$$



when $n > 0$, $h^{inv}[n] + a h^{inv}[n-1] = 0$,

or rewrite: $h^{inv}[n] = -a h^{inv}[n-1]$

$$h^{inv}[0] = 1$$

$$h^{inv}[1] = -a h^{inv}[0] = -a$$

$$h^{inv}[2] = -a h^{inv}[1] = a^2$$

$$h^{inv}[3] = -a h^{inv}[2] = -a^3$$

⋮

$$h^{inv}[n] = -a h^{inv}[n-1] = (-a)^n$$



The impulse response from the inverse system,

$$h^{inv}[n] = (-a)^n u[n]$$

Is the inverse system stable?

$$\sum_{k=0}^{\infty} |h^{inv}[k]| = \sum_{k=0}^{\infty} |a|^k$$

if $|a| < 1$, system stable; otherwise not stable.



Discrete-Time Step Response

當以單位步階訊號 Unit-Step 輸入時，系統輸出以 $s[n]$ 表示 **步階響應**：

$$s[n] = u[n] * h[n] = \sum_{k=-\infty}^{+\infty} h[k]u[n-k]$$

$$u[n-k] = \begin{cases} 1, & n-k \geq 0, \quad \text{or} \quad n \geq k \\ 0, & n-k < 0, \quad \text{or} \quad n < k \end{cases}$$



$$\therefore u[n-k] = \begin{cases} 1, & n \geq k \\ 0, & n < k \end{cases}$$

步階響應是脈衝響應的總和：

$$s[n] = \sum_{k=-\infty}^n h[k]$$

脈衝響應亦可用步階響應相減獲得：

$$h[n] = s[n] - s[n-1]$$



Continuous-Time Step Response

步階響應是脈衝響應的積分：

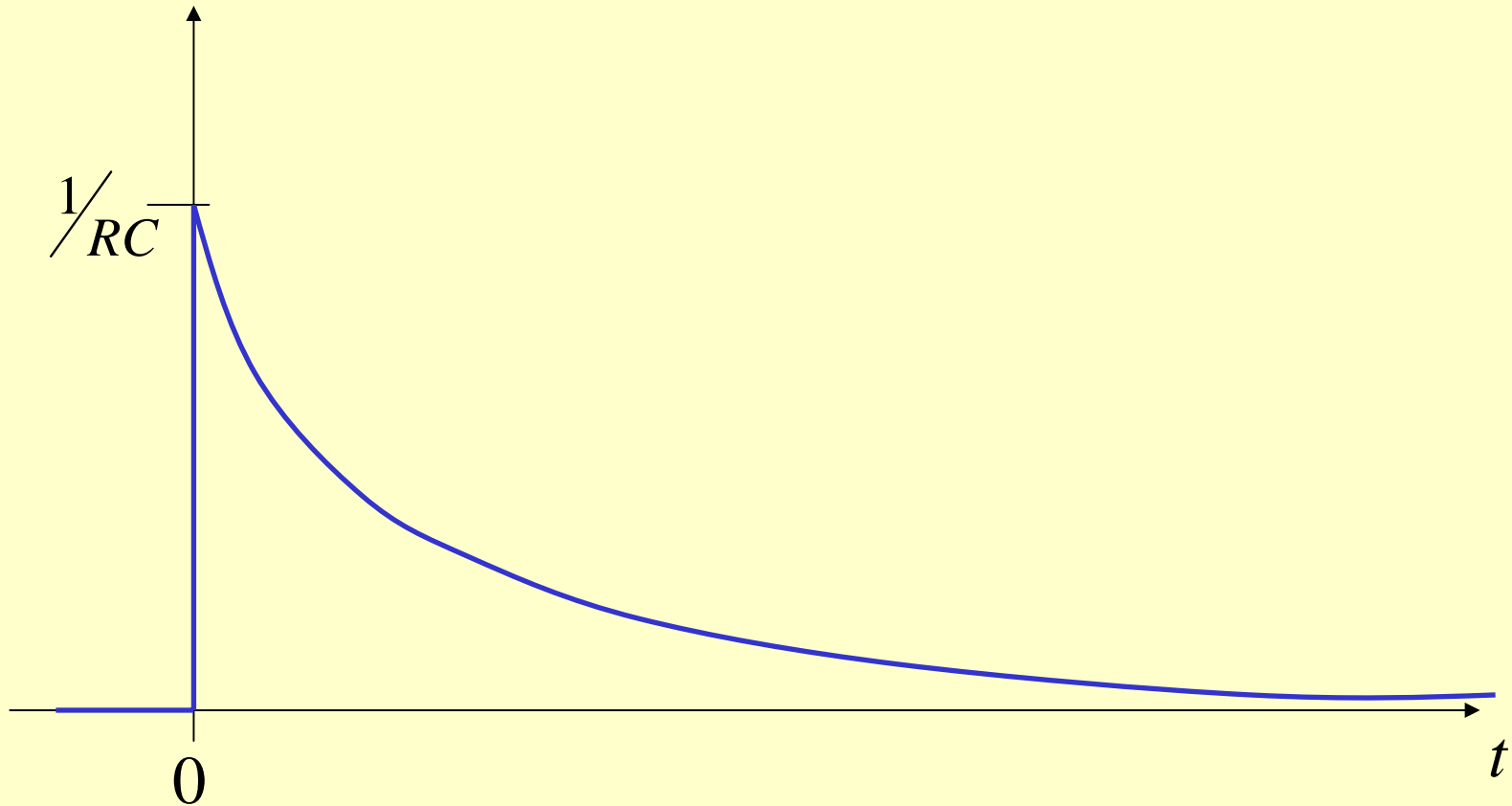
$$s(t) = \int_{-\infty}^t h(\tau) d\tau$$

脈衝響應亦可用步階響應微分獲得：

$$\frac{d}{dt} s(t) = \frac{d}{dt} \int_{-\infty}^t h(\tau) d\tau = h(t)$$



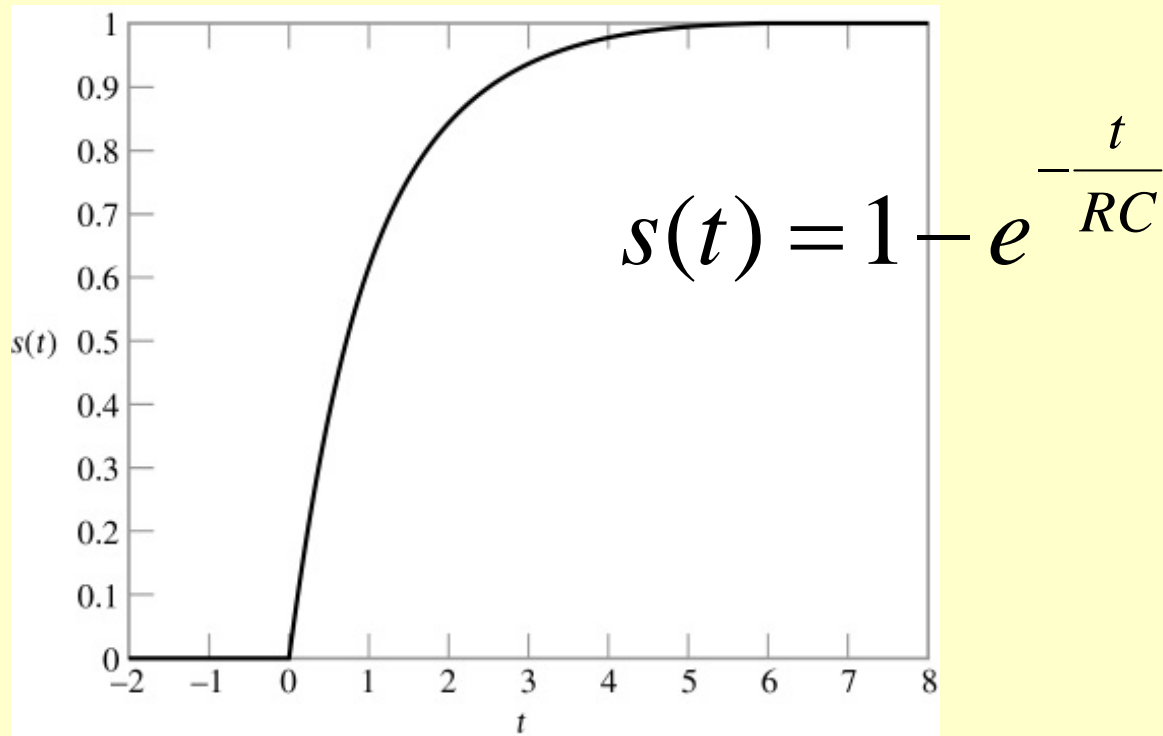
EX 2.14 已知脈衝響應， $h(t) = \frac{1}{RC} e^{-t/(RC)} u(t)$
找出步階響應 $s(t) = ?$





找出步階響應 = ?

Solution:



RC circuit step response for $RC = 1$ s



$$\therefore s(t) = \int_{-\infty}^t h(\tau) d\tau$$

$$s(t) = \int_{-\infty}^t \left[\frac{1}{RC} e^{-\frac{\tau}{RC}} u(\tau) \right] d\tau$$

$$= \frac{1}{RC} \int_0^t e^{-\frac{\tau}{RC}} d\tau = \left[-e^{-\frac{\tau}{RC}} \right] \Bigg|_0^t = -e^{-\frac{t}{RC}} + 1$$

$$= 1 - e^{-\frac{t}{RC}}$$



Differential and Difference Equation

微分方程式： 描述連續時間系統

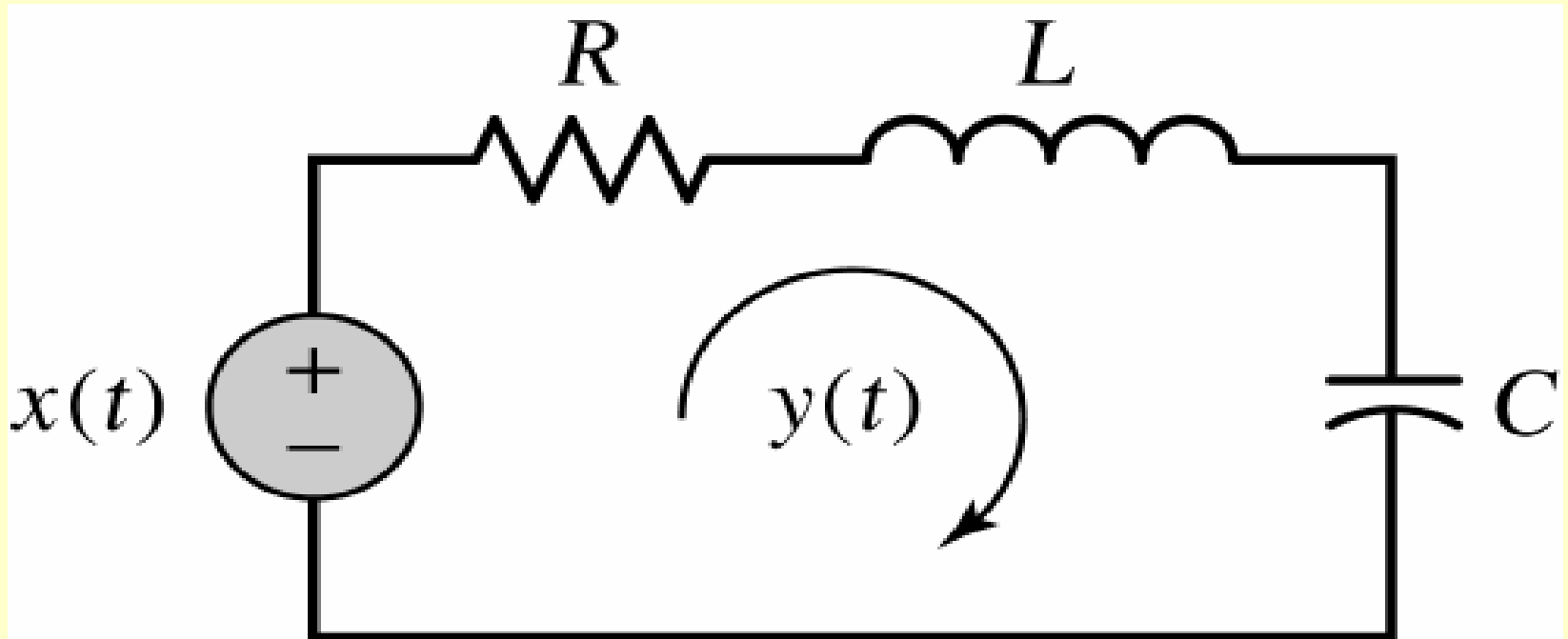
$$\sum_{k=0}^N a_k \frac{d^k}{dt^k} y(t) = \sum_{k=0}^M b_k \frac{d^k}{dt^k} x(t)$$

差分方程式： 描述離散時間系統

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$



Example of an *RLC* circuit described by a differential equation.





微分方程式：

輸入為電壓源： $x(t)$ 輸出為電流： $y(t)$

$$L \frac{d}{dt} y(t) + Ry(t) + \frac{1}{C} \int_{-\infty}^t y(\tau) d\tau = x(t)$$

對上式等號兩邊對t微分：

$$L \frac{d^2}{dt^2} y(t) + R \frac{d}{dt} y(t) + \frac{1}{C} y(t) = \frac{d}{dt} x(t)$$



差分方程式：

$$\sum_{k=0}^N a_k y[n-k] = \sum_{k=0}^M b_k x[n-k]$$

$$y[n] = \frac{1}{a_0} \sum_{k=0}^M b_k x[n-k] - \frac{1}{a_0} \sum_{k=1}^N a_k y[n-k]$$



EX: 2.16 Difference Equation:

$$y[n] - 1.143y[n-1] + 0.4128y[n-2] = \\ 0.0675x[n] + 0.1349x[n-1] + 0.675x[n-2]$$

Initial Condition: $x[n] = 0$, $y[-1] = 1$, and $y[-2] = 2$,
Please write a recursive rule to find $y[n] = ?$



$$y[n] = 1.143y[n-1] - 0.4128y[n-2] + 0.0675x[n] + 0.1349x[n-1] + 0.675x[n-2]$$

Solution:

$$\because x[n] = 0, \quad \therefore y[n] = 1.143y[n-1] - 0.4128y[n-2]$$

$$y[-2] = 2 \quad \leftarrow \text{已知}$$

$$y[-1] = 1 \quad \leftarrow$$

$$y[0] = 1.143(1) - 0.4128(2) = 0.3174$$

$$y[1] = 1.143(0.3174) - 0.4128(1) = 0.3628 - 0.4128 = -0.05$$

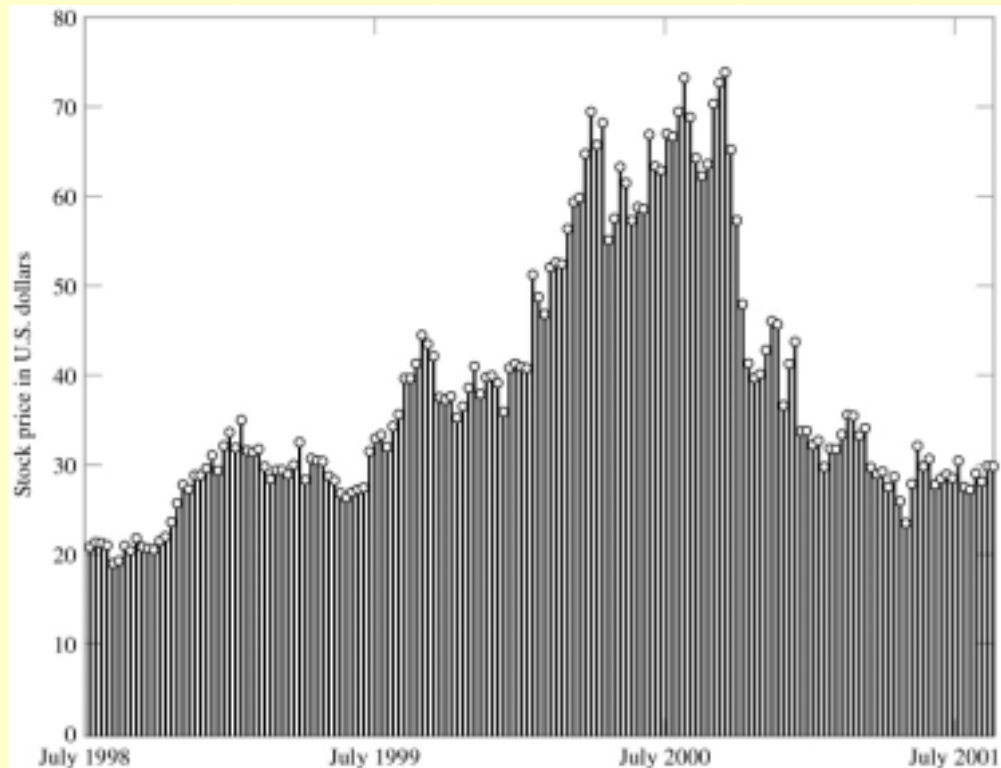
$$y[2] = 1.143(-0.05) - 0.4128(0.3174) = -0.0572 - 0.1311 = -0.1883$$

$$y[3] = 1.143(-0.1883) - 0.4128(-0.05) = -0.0406 + 0.0207 = -0.0199$$

⋮



EX: 2.16 : If the input $x[n]$ is described by the following closing price of INTEL stock, please find the output $y[n] = ?$



Weekly closing price of Intel stock



Output associated with the weekly closing price of Intel stock. (討論輸出、入差異)

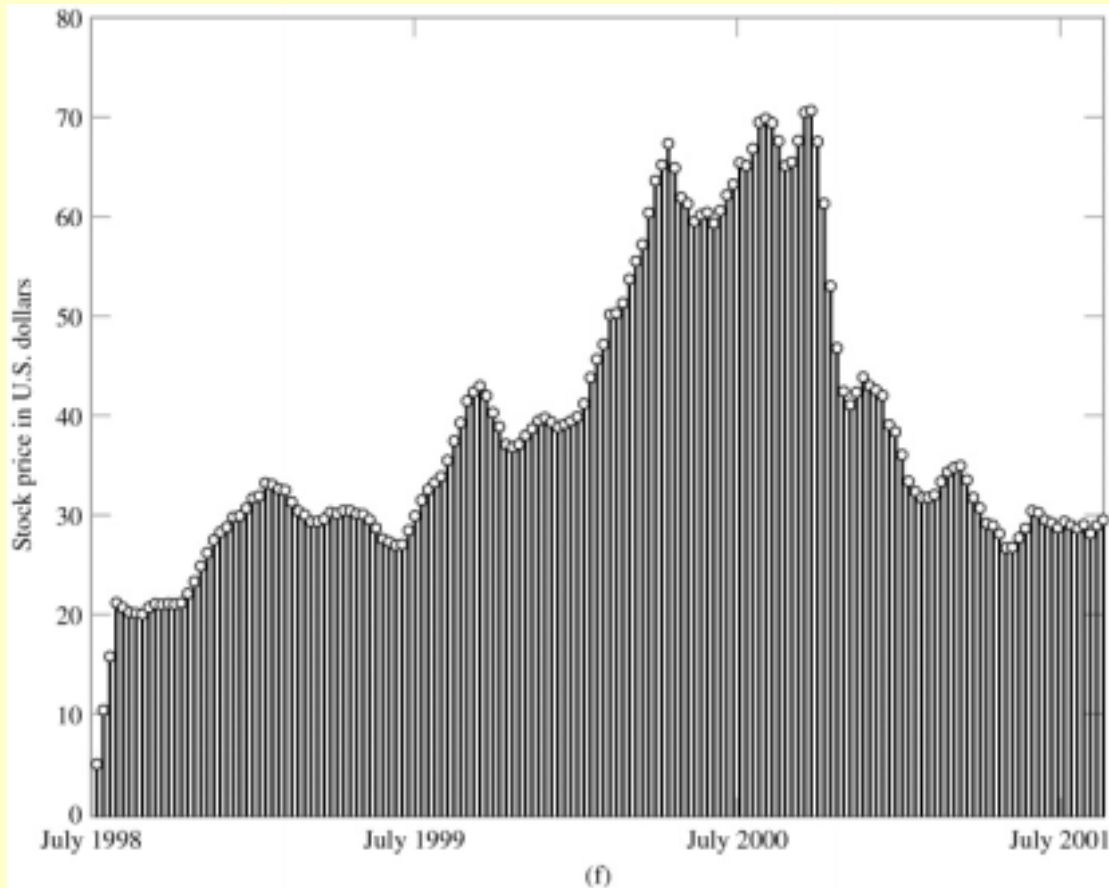
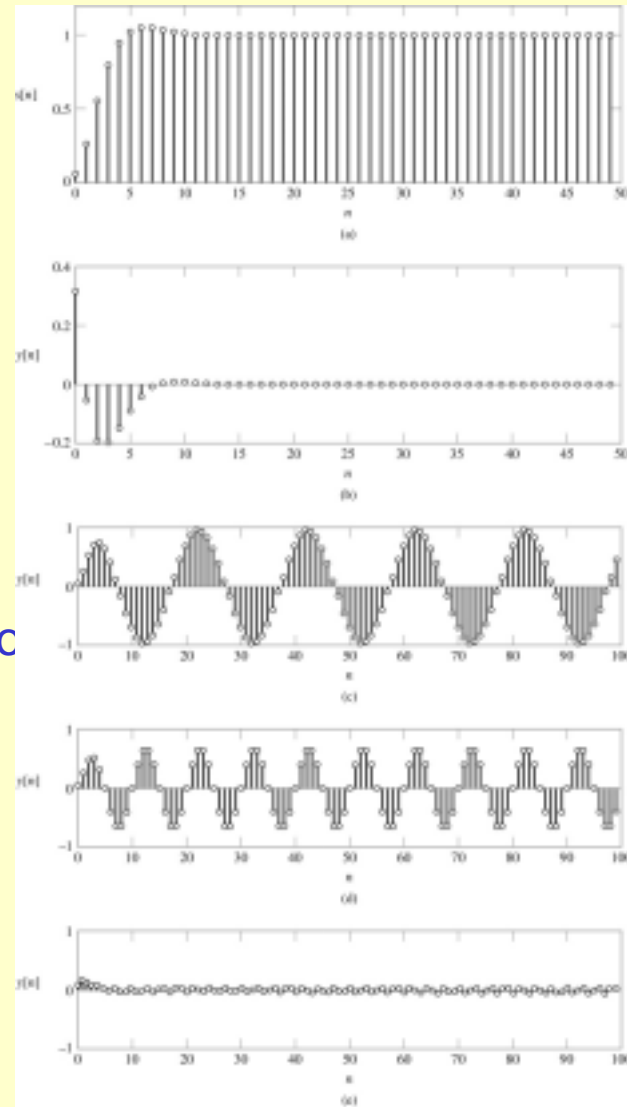




Illustration of the solution to Example 2.16.

- (a) Step response of system.
- (b) Output due to nonzero initial conditions with zero input.
- (c) Output due to $x_1[n] = \cos(1/10\pi n)$.
- (d) Output due to $x_2[n] = \cos(1/5\pi n)$.
- (e) Output due to $x_3[n] = \cos(7/10\pi n)$.





EX2.14 請用微分方程式描述下述 RL circuit , 其 $x(t)$ 表輸入電壓、 $y(t)$ 表輸出電流。

學生試一試 ?

